

ECE4270 Fundamentals of DSP

Lecture 3 Frequency Response of LTI Systems

School of Electrical and Computer
Engineering
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Overview of Lecture 3

- Announcement
- Difference equations (from Last Lecture Slides)
- Initial rest conditions and LTI (from Last Lecture)
- IIR systems
- FIR systems
- Matlab and LTI systems
- Complex exponential inputs to LTI systems
- The frequency response
- Examples
 - Delay, First difference, Moving average
- Plotting the frequency response

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Announcement

- TA information:
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 - office hours:
 - Thursday 9-10
 - MW 4:30 PM-5:30 PM
 - Location: VL C449. Fourth floor of Van Leer Building, Central corridor.

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Difference Equations

- For all computationally realizable LTI systems, the input and output satisfy a difference equation of the form

$$\sum_{k=0}^N a_k y[n-k] = \sum_{k=0}^M b_k x[n-k]$$

- This leads to the recurrence formula

$$y[n] = -\sum_{k=1}^N \left(\frac{a_k}{a_0}\right) y[n-k] + \sum_{k=0}^M \left(\frac{b_k}{a_0}\right) x[n-k]$$

which can be used to compute the “present” output from the present and M past values of the input and N past values of the output

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Recursive Computation of Output

Let $x[n] = K\delta[n]$ and $y[-1] = c$.

n	$x[n]$	$y[n] = ay[n-1] + x[n]$
...
-1	0	c
0	K	$ac + K$
1	0	$a(ac + K) = a^2c + Ka$
2	0	$a(a^2c + Ka) = a^3c + Ka^2$
3	0	$a(a^3c + Ka^2) = a^4c + Ka^3$
...

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General Solution

- By induction, we see that if

$$y[n] = ay[n-1] + x[n]$$

with $x[n] = K\delta[n]$ and $y[-1] = c$.

then the solution is

$$y[n] = ca^{n+1} + Ka^n \text{ for } n \geq -1$$

- If $x[n] = K\delta[n - n_d]$, then the output will be

$$y[n] = ca^{n+1} + Ka^{n-n_d} \text{ for } n \geq -1$$

Implies
Not TI

- If $x[n] = \delta[n]$, then the output will be

$$y[n] = ca^{n+1} + a^n \text{ for } n \geq -1$$

Implies
Not linear

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LTI Recursive Implementation

- We say that an input is *suddenly applied at time n_d* if $x[n]=0$ for all $n < n_d$.
- If the input is suddenly applied and we assume that $y[n]=0$ for all $n < n_d$, then the iterative computation will be both linear and time-invariant. This assumption provides the required set of **auxiliary conditions** $\{y[n_d-1], y[n_d-2], \dots, y[n_d-N]\}$ that is required to get the recursion going. Zero auxiliary conditions are called the **initial rest conditions**.
- For the first-order case of our example, the impulse response of the LTI system is

$$h[n] = a^n u[n]$$

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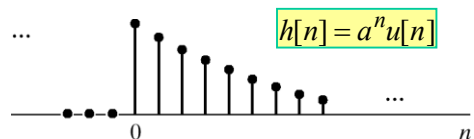
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Exponential Impulse Response

- With initial rest conditions, the difference equation

$$y[n] = ay[n-1] + x[n]$$

has impulse response



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LTI Recursion of First-Order DE

- If we assume initial rest conditions, then the difference equation $y[n] = ay[n-1] + x[n]$

has impulse response $h[n] = a^n u[n]$

- In other words, $y[n]$ is also given by the convolution

$$y[n] = \sum_{k=-\infty}^{\infty} x[k] a^{n-k} u[n-k] = \sum_{k=-\infty}^n x[k] a^{n-k}$$

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IIR Systems

- Under conditions of initial rest, a system whose input and output satisfy a difference equation of the form

$$y[n] = \underbrace{\sum_{k=1}^N a_k y[n-k]}_{\text{feedback}} + \underbrace{\sum_{k=0}^M b_k x[n-k]}_{\text{feedforward}}$$

Note the redefinition of the coefficients.

is LTI and its impulse response is of the form

$$h[n] = \sum_{k=1}^N A_k \alpha_k^n u[n] = \begin{cases} A_k \alpha_k^n, & n \geq 0 \\ 0, & n < 0 \end{cases}$$

⇒ Infinite duration Impulse Response (IIR)

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FIR Systems

- If there are no feedback terms, then the difference equation becomes

$$y[n] = \sum_{k=0}^M b_k x[n-k]$$

- This system is LTI and its impulse response is

$$h[n] = \sum_{k=0}^M b_k \delta[n-k] = \begin{cases} b_n, & 0 \leq n \leq M \\ 0, & \text{otherwise} \end{cases}$$

⇒ Finite duration Impulse Response (FIR)

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Moving Average System

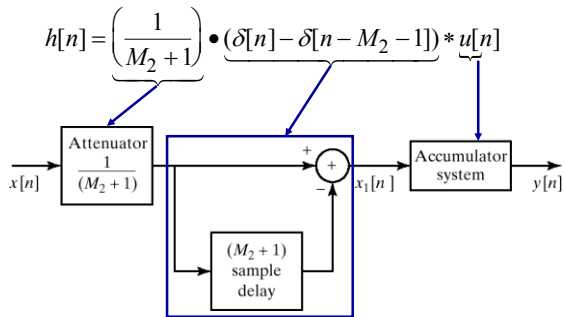
$$y[n] = \frac{1}{M_2 + 1} \sum_{k=0}^{M_2} x[n-k]$$

$$\begin{aligned} h[n] &= \frac{1}{M_2 + 1} \sum_{k=0}^{M_2} \delta[n-k] = \frac{1}{M_2 + 1} \begin{cases} 1 & 0 \leq n \leq M_2 \\ 0 & \text{otherwise} \end{cases} \\ &= \frac{1}{M_2 + 1} (u[n] - u[n - M_2 - 1]) \\ &= \frac{1}{M_2 + 1} (\delta[n] - \delta[n - M_2 - 1]) * u[n] \end{aligned}$$

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Moving Average System

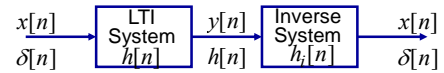


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Inverse Systems

- An inverse system compensates (undoes) the effects of another system.



$$\Rightarrow h[n] * h_i[n] = \delta[n]$$

- The accumulator and first-difference systems are inverses of each other.

$$(\delta[n] - \delta[n - 1]) * u[n] = u[n] - u[n - 1] = \delta[n]$$

- Understanding inverse systems is greatly facilitated by transform methods.

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MATLAB and LTI Systems

»help conv

CONV Convolution and polynomial multiplication.

$Y = \text{CONV}(X, H)$ convolves vectors X and H . The resulting vector is length $\text{LENGTH}(X) + \text{LENGTH}(H) - 1$.

If X and H are vectors of polynomial coefficients, convolving them is equivalent to multiplying the two polynomials.

»help filter

FILTER One-dimensional digital filter.

$Y = \text{FILTER}(B, A, X)$ filters the data in vector X with the filter described by vectors A and B to create the filtered data Y . The filter is a "Direct Form II Transposed" implementation of the standard difference equation:

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MATLAB and LTI Systems

- The moving average system

```
>> h=ones(1,M+1)/(M+1);
```

```
>> y=conv(x,h);
```

- The accumulator system:

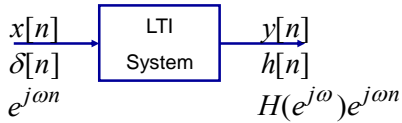
```
>> b=1; a=[1,-1];
```

```
>> y=filter(b,a,x);
```

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Complex Exponential Input Signals



$$y[n] = \sum_{k=-\infty}^{\infty} x[k]h[n-k] = \sum_{k=-\infty}^{\infty} h[k]x[n-k]$$

$$y[n] = \sum_{k=-\infty}^{\infty} h[k]e^{j\omega(n-k)} = \sum_{k=-\infty}^{\infty} h[k]e^{-j\omega k} e^{j\omega n}$$

$$= H(e^{j\omega})e^{j\omega n}$$

Frequency response

$$H(e^{j\omega}) = \sum_{k=-\infty}^{\infty} h[k]e^{-j\omega k}$$

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The Frequency Response

$$H(e^{j\omega}) = \sum_{k=-\infty}^{\infty} h[k]e^{-j\omega k}$$

- Periodicity of the frequency response

$$H(e^{j(\omega+2\pi)}) = \sum_{k=-\infty}^{\infty} h[k]e^{-j(\omega+2\pi)k} = \sum_{k=-\infty}^{\infty} h[k]e^{-j\omega k} e^{-j2\pi k}$$

- Existence of the frequency response

$$|H(e^{j\omega})| = \left| \sum_{k=-\infty}^{\infty} h[k]e^{-j\omega k} \right| \leq \sum_{k=-\infty}^{\infty} |h[k]| |e^{-j\omega k}|$$

$$|H(e^{j\omega})| \leq \sum_{k=-\infty}^{\infty} |h[k]| < \infty$$

Same as condition for stability!

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Examples

- Delay: $y[n] = x[n - n_d]$

$$h[n] = \delta[n - n_d] \Rightarrow x[n] * \delta[n - n_d] = x[n - n_d]$$

- Moving Average: $y[n] = \frac{1}{M+1} \sum_{k=0}^M x[n-k]$

$$h[n] = \frac{1}{M+1} \sum_{k=0}^M \delta[n-k]$$

- First difference: $y[n] = x[n] - x[n-1]$

$$h[n] = \delta[n] - \delta[n-1]$$

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Delay and First Difference

- Delay: $y[n] = x[n - n_d]$ $h[n] = \delta[n - n_d]$

$$x[n] = e^{j\omega n} \mapsto y[n] = e^{j\omega(n-n_d)} = \underbrace{e^{-j\omega n_d}}_{H(e^{j\omega})} e^{j\omega n}$$

$$H(e^{j\omega}) = e^{-j\omega n_d}$$

- First difference:

$$y[n] = x[n] - x[n-1] \quad h[n] = \delta[n] - \delta[n-1]$$

$$x[n] = e^{j\omega n} \mapsto y[n] = e^{j\omega n} - e^{j\omega(n-1)} = \underbrace{(1 - e^{-j\omega})}_{H(e^{j\omega})} e^{j\omega n}$$

$$H(e^{j\omega}) = 1 - e^{-j\omega}$$

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Moving Average

$$y[n] = \frac{1}{M+1} \sum_{k=0}^M x[n-k]$$

$$\begin{aligned} x[n] = e^{j\omega n} \mapsto y[n] &= \frac{1}{M+1} \sum_{k=0}^M e^{j\omega(n-k)} \\ &= \frac{1}{M+1} \sum_{k=0}^M e^{-j\omega k} e^{j\omega n} \\ &= \underbrace{\frac{1}{M+1} \sum_{k=0}^M e^{-j\omega k}}_{H(e^{j\omega})} e^{j\omega n} \end{aligned}$$

$$H(e^{j\omega}) = \frac{1}{M+1} \sum_{k=0}^M e^{-j\omega k}$$

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Moving Average Frequency Response

$$H(e^{j\omega}) = \frac{1}{M+1} \sum_{k=0}^M e^{-j\omega k} = \frac{1}{M+1} \frac{(1 - e^{-j\omega(M+1)})}{(1 - e^{-j\omega})}$$

$$H(e^{j\omega}) = \frac{1}{M+1} \frac{(e^{j\omega(M+1)/2} - e^{-j\omega(M+1)/2})e^{-j\omega(M+1)/2}}{(e^{j\omega/2} - e^{-j\omega/2})e^{-j\omega/2}}$$

$$H(e^{j\omega}) = \frac{1}{M+1} \frac{\sin[\omega(M+1)/2]}{\sin(\omega/2)} e^{-j\omega M/2}$$

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Plotting using MATLAB

»help freqz

FREQZ Z-transform digital filter frequency response.

When N is an integer, [H,W] = FREQZ(B,A,N) returns the N-point frequency vector W in radians and the N-point complex frequency response vector H given numerator and denominator coefficients in vectors B and A. The frequency response is evaluated at N points equally spaced around the upper half of the unit circle. If N isn't specified, it defaults to 512.

```
>> omega=(0:400)*pi/400; b=[1,1,1,1,1]/5;
```

```
>> H=freqz(b,1,omega);
```

```
>> subplot(211); plot(omega/pi,abs(H))
```

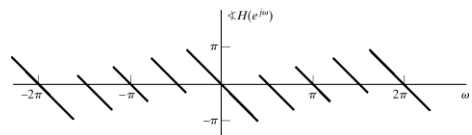
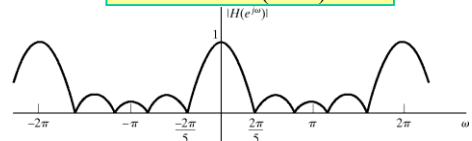
```
>> subplot(212); plot(omega/pi,angle(H))
```

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Plotting the Frequency Response

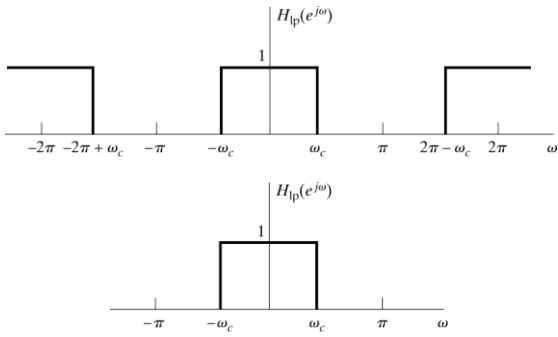
$$H(e^{j\omega}) = \frac{1}{5} \frac{\sin[\omega 5/2]}{\sin(\omega/2)} e^{-j\omega 2} \quad M=4$$



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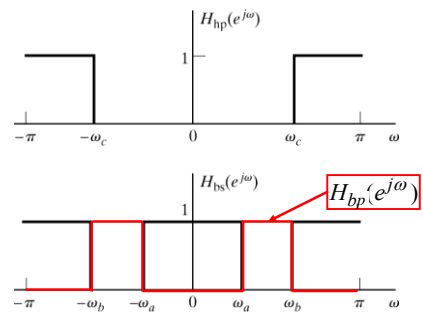
Ideal Lowpass Filter



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Frequency Selective Filters



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