

**ECE4270**  
**Fundamentals of DSP**

**Lecture 11**

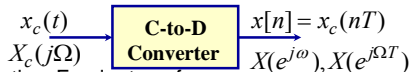
**Sampling and Reconstruction from Samples**

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Center for Signal and Information Processing  
Georgia Institute of Technology

**Overview of Lecture**

- Chapter 4--Sampling
  - Bandlimited signals
  - Ideal Continuous-to-Discrete conversion (sampling)
  - Derivation of frequency-domain formula
  - Interpretation of frequency-domain formula
    - Oversampling
    - Aliasing distortion
- Band Limited Reconstruction

**Sampling (C-to-D Conversion)**



- Discrete-time Fourier transform:

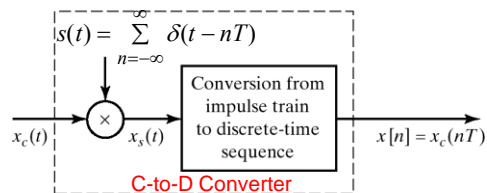
$$X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x[n]e^{-j\omega n}$$

- Frequency-domain relation:

$$X(e^{j\Omega T}) = \sum_{n=-\infty}^{\infty} x[n]e^{-j\Omega T n} = \frac{1}{T} \sum_{k=-\infty}^{\infty} X_c(j(\Omega - k\Omega_s))$$

- Sampling frequency:  $\Omega_s = 2\pi/T$
- Normalized frequency:  $\omega = \Omega T$

**Derivation of Basic FT Formula - I**



$$x_s(t) = x_c(t) \cdot \sum_{n=-\infty}^{\infty} \delta(t - nT) = \sum_{n=-\infty}^{\infty} x_c(t)\delta(t - nT)$$

$$x_s(t) = \sum_{n=-\infty}^{\infty} x_c(nT)\delta(t - nT) = \sum_{n=-\infty}^{\infty} x[n]\delta(t - nT)$$

## Derivation of Basic FT Formula - II

$$x_s(t) = x_c(t) \cdot s(t) = x_c(t) \sum_{n=-\infty}^{\infty} \delta(t - nT)$$

$$s(t) = \sum_{n=-\infty}^{\infty} \delta(t - nT) \Leftrightarrow S(j\Omega) = \sum_{k=-\infty}^{\infty} \frac{2\pi}{T} \delta(\Omega - k \frac{2\pi}{T})$$

$$X_s(j\Omega) = \frac{1}{2\pi} X_c(j\Omega) * S(j\Omega)$$

$$X_s(j\Omega) = \frac{1}{2\pi} X_c(j\Omega) * \sum_{k=-\infty}^{\infty} \frac{2\pi}{T} \delta(\Omega - k \frac{2\pi}{T})$$

$$X_s(j\Omega) = \frac{1}{T} \sum_{k=-\infty}^{\infty} X_c(\Omega - k \frac{2\pi}{T})$$

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## Derivation of Basic FT Formula - III

$$x_s(t) = x_c(t) \sum_{n=-\infty}^{\infty} \delta(t - nT) = \sum_{n=-\infty}^{\infty} x[n] \delta(t - nT)$$

$$X_s(j\Omega) = \sum_{n=-\infty}^{\infty} x[n] e^{-j\Omega n T}$$

$$X_s(j\Omega) = \sum_{n=-\infty}^{\infty} x[n] e^{-j(\Omega T)n} = X(e^{j\Omega T})$$

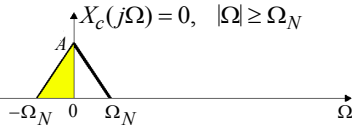
$$X(e^{j\Omega T}) = \frac{1}{T} \sum_{k=-\infty}^{\infty} X_c(\Omega - k \frac{2\pi}{T})$$

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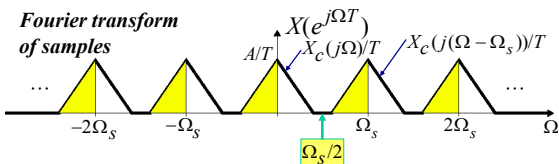
## Oversampling

“Typical”  
bandlimited signal



$$X(e^{j\Omega T}) = \frac{1}{T} \sum_{k=-\infty}^{\infty} X_c(j(\Omega - k\Omega_s)) \quad \Omega_s = 2\pi/T$$

Fourier transform  
of samples

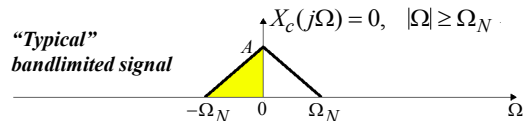


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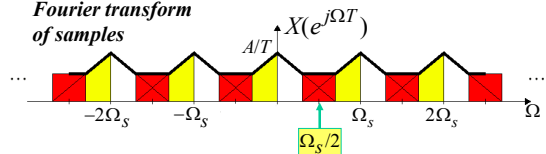
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## Undersampling (Aliasing Distortion)

- If  $\Omega_s < 2\Omega_N$ , the copies of  $X_c(j\Omega)$  overlap, and we have **aliasing distortion**.



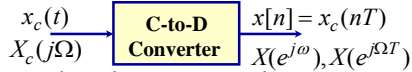
Fourier transform  
of samples



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## Sampling Theorem



- Frequency-domain representation

$$X(e^{jΩT}) = \sum_{n=-\infty}^{\infty} x[n]e^{-jΩTn} = \frac{1}{T} \sum_{k=-\infty}^{\infty} X_c(j(\Omega - k\Omega_s))$$

- Sampling theorem justification:

If  $X_c(j\Omega) = 0$ ,  $|\Omega| \geq \Omega_N$  and  $\Omega_s/2 = \pi/T \geq \Omega_N$ ,

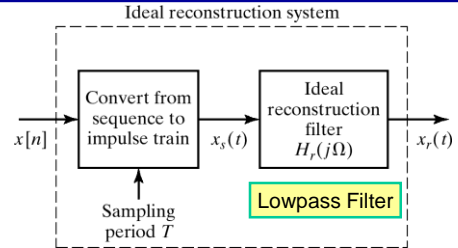
- Therefore we should be able to recover  $x_c(t)$ !

then  $X(e^{jΩT}) = \frac{1}{T} X_c(j\Omega)$ ,  $|\Omega| \leq \pi/T = \Omega_s/2$

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## Reconstruction from Samples (I)



$$x_r(t) = x_s(t) * h_r(t) = \sum_{n=-\infty}^{\infty} x[n]\delta(t - nT) * h_r(t)$$

$$x_r(t) = \sum_{n=-\infty}^{\infty} x[n]h_r(t - nT)$$

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## Reconstruction from Samples (II)



$$x_r(t) = \sum_{n=-\infty}^{\infty} x[n]h_r(t - nT)$$

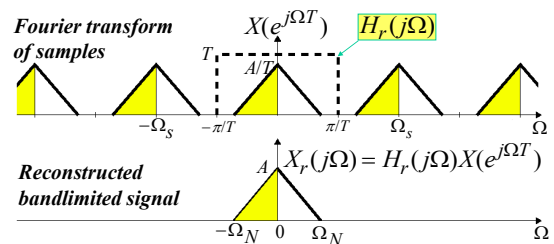
- Frequency-domain representation

$$\begin{aligned} X_r(j\Omega) &= \sum_{n=-\infty}^{\infty} x[n] \left( e^{-j\Omega T n} H_r(j\Omega) \right) \\ &= \left( \sum_{n=-\infty}^{\infty} x[n] e^{-j\Omega T n} \right) H_r(j\Omega) = X(e^{j\Omega T}) H_r(j\Omega) \\ &= \left( \frac{1}{T} \sum_{k=-\infty}^{\infty} X_c(j(\Omega - 2\pi k / T)) \right) H_r(j\Omega) \end{aligned}$$

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## Bandlimited Reconstruction (I)



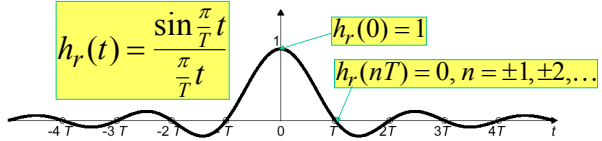
- If  $\Omega_s \geq 2\Omega_N$ , the copies of  $X_c(j\Omega)$  do not overlap, so  $X_r(j\Omega) = H_r(j\Omega) X(e^{j\Omega T}) = X_c(j\Omega)$ , and we get perfect reconstruction.

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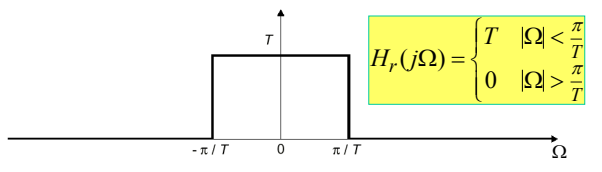
### Bandlimited Reconstruction (II): Reconstruction Filter

$$h_r(t) = \frac{\sin \frac{\pi}{T} t}{\frac{\pi}{T} t}$$



$$h_r(0) = 1$$

$$h_r(nT) = 0, n = \pm 1, \pm 2, \dots$$



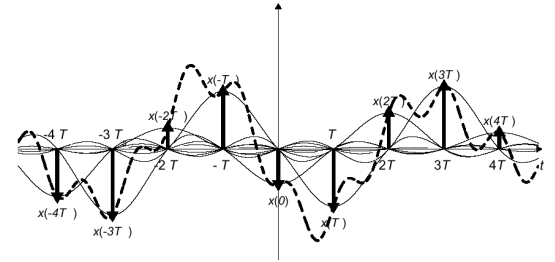
$$H_r(j\Omega) = \begin{cases} T & |\Omega| < \frac{\pi}{T} \\ 0 & |\Omega| > \frac{\pi}{T} \end{cases}$$

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### Bandlimited Reconstruction (III): (Time-Domain)

$$x_r(t) = \sum_{n=-\infty}^{\infty} x(nT)h_r(t-nT) = \sum_{n=-\infty}^{\infty} x(nT) \frac{\sin[\pi(t-nT)/T]}{\pi(t-nT)/T}$$



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